

BUH515D

HIGH VOLTAGE FAST-SWITCHING NPN POWER TRANSISTOR

- SGS-THOMSON PREFERRED SALESTYPE
- HIGH VOLTAGE CAPABILITY
- U.L. RECOGNISED ISOWATT218 PACKAGE (U.L. FILE # E81734 (N))
- NPN TRANSISTOR WITH INTEGRATED FREEWHEELING DIODE

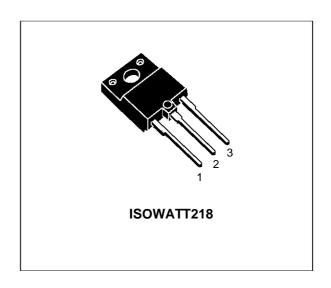
APPLICATIONS:

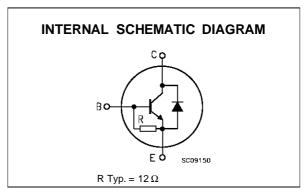
 HORIZONTAL DEFLECTION FOR COLOUR TV

DESCRIPTION

The BUH515D is manufactured using Multiepitaxial Mesa technology for cost-effective high performance and uses a Hollow Emitter structure to enhance switching speeds.

The BUH series is designed for use in horizontal deflection circuits in televisions and monitors.





ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V_{CBO}	Collector-Base Voltage (I _E = 0)	1500	V
V_{CEO}	Collector-Emitter Voltage (I _B = 0)	700	V
V_{EBO}	Emitter-Base Voltage (I _C = 0)	5	V
Ic	Collector Current	8	Α
Ісм	Collector Peak Current (tp < 5 ms)	15	Α
lΒ	Base Current	5	Α
I _{BM}	Base Peak Current (t _p < 5 ms)	8	Α
P _{tot}	Total Dissipation at T _c = 25 °C	50	W
T_{stg}	Storage Temperature	-65 to 150	°C
Tj	Max. Operating Junction Temperature	150	°C

August 1996 1/7

THERMAL DATA

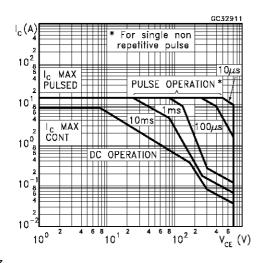
R _{thj-case} Thermal Resistance Junction-case	Max 2.5	°C/W	
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ELECTRICAL CHARACTERISTICS (T_{case} = 25 °C unless otherwise specified)

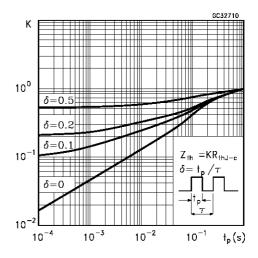
Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
I _{CES}	Collector Cut-off Current (V _{BE} = 0)	$V_{CE} = 1300 \text{ V}$ $V_{CE} = 1500 \text{ V}$ $V_{CE} = 1500 \text{ V}$ $T_j = 125 \text{ °C}$			10 1 2	μA mA mA
I _{EBO}	Emitter Cut-off Current (I _C = 0)	V _{EB} = 5 V			200	mA
V _{CE(sat)} *	Collector-Emitter Saturation Voltage	I _C = 5 A I _B = 1.25 A			1.5	V
$V_{BE(sat)^*}$	Base-Emitter Saturation Voltage	I _C = 5 A I _B = 1.25 A			1.3	V
h _{FE} *	DC Current Gain	I _C = 5 A V _{CE} = 5 V I _C = 5 A V _{CE} = 5 V T _j = 100 °C	5 3		10	
t _s t _f	RESISTIVE LOAD Storage Time Fall Time	$V_{CC} = 400 \text{ V}$ $I_{C} = 5 \text{ A}$ $I_{B1} = 1.5 \text{ A}$ $I_{B2} = -2.5 \text{ A}$		2.4 170	3.6 260	μs ns
t _s t _f	INDUCTIVE LOAD Storage Time Fall Time	$I_{C} = 5 \text{ A}$ $f = 15625 \text{ Hz}$ $I_{B1} = 1.25 \text{ A}$ $I_{B2} = -2.5 \text{ A}$ $V_{ceflyback} = 1050 \sin\left(\frac{\pi}{10} \cdot 10^{6}\right) t$ V		3.5 450		μs ns
V _F	Diode Forward Voltage	I _F = 5 A			2	V

^{*} Pulsed: Pulse duration = 300 μs, duty cycle 1.5 %

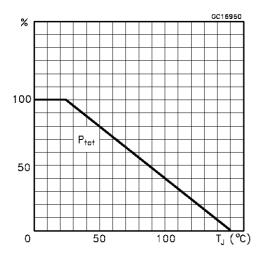
Safe Operating Area



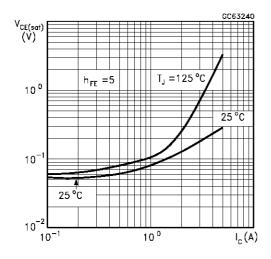
Thermal Impedance



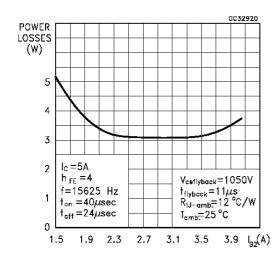
Derating Curve



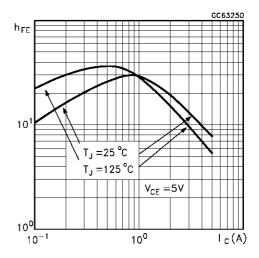
Collector Emitter Saturation Voltage



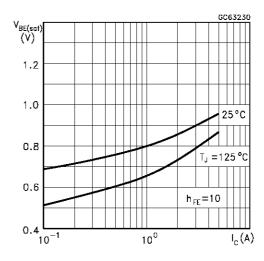
Power Losses at 16 KHz



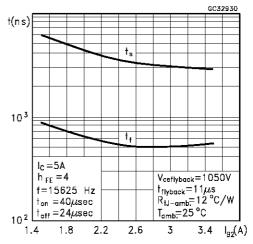
DC Current Gain



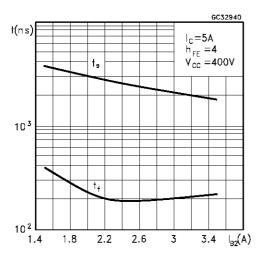
Base Emitter Saturation Voltage



Switching Time Inductive Load at 16KHz (see figure 2)



Switching Time Resistive Load



BASE DRIVE INFORMATION

In order to saturate the power switch and reduce conduction losses, adequate direct base current I_{B1} has to be provided for the lowest gain I_{FE} at 100 I_{C} (line scan phase). On the other hand, negative base current I_{B2} must be provided to turn off the power transistor (retrace phase). Most of the dissipation, especially in the deflection application, occurs at switch-off. Therefore it is essential to determine the value of I_{B2} which minimizes power losses, fall time I_{f} and, consequently, I_{f} . A new set of curves have been defined to give total power losses, I_{S} and I_{f} as a function of I_{B2} at 16 KHz frequencies for choosing the optimum negative drive. The test circuit is illustrated in fig. 1.

Inductance L_1 serves to control the slope of the negative base current I_{B2} to recombine the excess carrier in the collector when base current is still present, this avoid any tailing phenomenon in the collector current.

The values of L and C are calculated from the following equations:

$$\frac{1}{2}L(I_C)^2 = \frac{1}{2}C(V_{CEfly})^2$$
$$\omega = 2\pi f = \frac{1}{\sqrt{LC}}$$

Where I_{C} = operating collector current, V_{CEfly} = flyback voltage, f= frequency of oscillation during retrace.

Figure 1: Inductive Load Switching Test Circuit

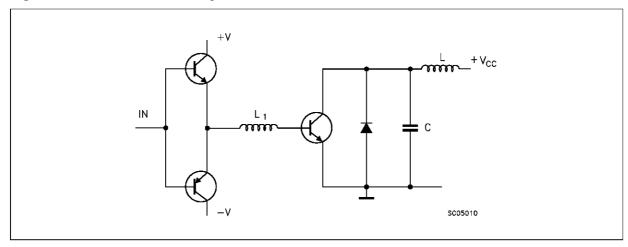
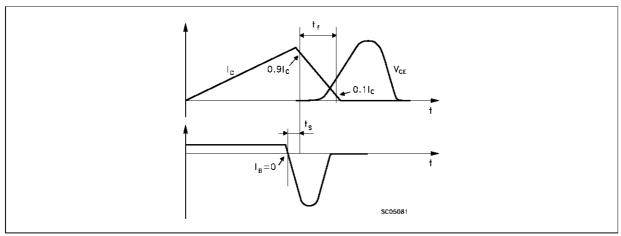
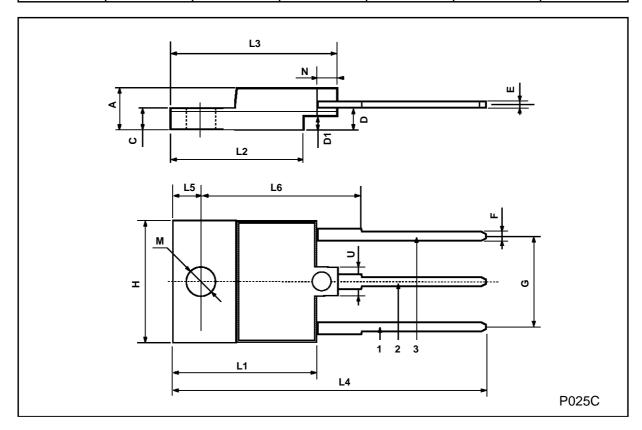


Figure 2: Switching Waveforms in a Deflection Circuit



ISOWATT218 MECHANICAL DATA

DIM.	mm		inch			
DIIVI.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
А	5.35		5.65	0.210		0.222
С	3.3		3.8	0.130		0.149
D	2.9		3.1	0.114		0.122
D1	1.88		2.08	0.074		0.081
Е	0.75		1	0.029		0.039
F	1.05		1.25	0.041		0.049
G	10.8		11.2	0.425		0.441
Н	15.8		16.2	0.622		0.637
L1	20.8		21.2	0.818		0.834
L2	19.1		19.9	0.752		0.783
L3	22.8		23.6	0.897		0.929
L4	40.5		42.5	1.594		1.673
L5	4.85		5.25	0.190		0.206
L6	20.25		20.75	0.797		0.817
М	3.5		3.7	0.137		0.145
N	2.1		2.3	0.082		0.090
U		4.6			0.181	



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