

BU808DFI

HIGH VOLTAGE FAST-SWITCHING NPN POWER DARLINGTON

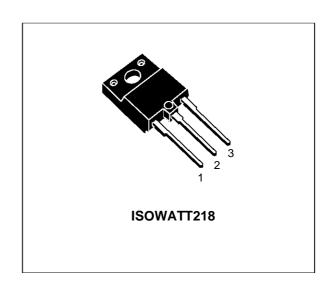
- SGS-THOMSON PREFERRED SALESTYPE
- NPN DARLINGTON
- HIGH VOLTAGE CAPABILITY
- HIGH DC CURRENT GAIN
- U.L. RECOGNISED ISOWATT218 PACKAGE (U.L. FILE # E81734 (N))
- LOW BASE-DRIVE REQUIREMENTS.

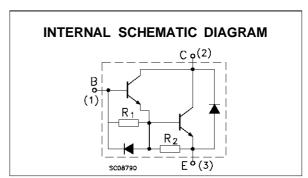
APPLICATIONS

 HORIZONTAL DEFLECTION FOR COLOUR TV

DESCRIPTION

The BU808DFI is manufactured using Multiepitaxial Mesa technology for cost-effective high performance.





ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V_{CBO}	Collector-Base Voltage (I _E = 0)	1400	V
V _{CEO}	Collector-Emitter Voltage (I _B = 0)	700	V
V _{EBO}	Emitter-Base Voltage (I _C = 0)	5	V
Ic	Collector Current	8	A
I _{CM}	Collector Peak Current (t _p < 5 ms)	10	A
I _B	Base Current	3	A
I _{BM}	Base Peak Current (t _p < 5 ms)	6	A
P _{tot}	Total Dissipation at T _c = 25 °C	52	W
T _{stg}	Storage Temperature	-65 to 150	°C
Tj	Max. Operating Junction Temperature	150	°C

June 1996 1/7

BU808DFI

THERMAL DATA

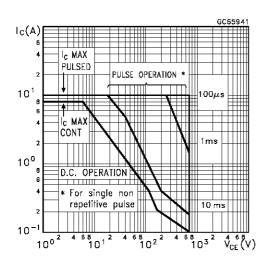
R _{thj-case} Thermal Res	stance Junction-case	Max	2.4	°C/W	
-----------------------------------	----------------------	-----	-----	------	--

ELECTRICAL CHARACTERISTICS (T_{case} = 25 °C unless otherwise specified)

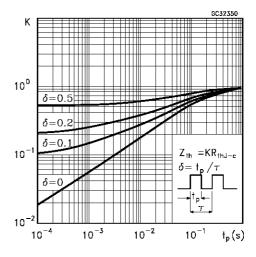
Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
I _{CES}	Collector Cut-off Current (V _{BE} = 0)	V _{CE} = 1400 V			400	μΑ
I _{CEX}	Collector Cut-off Current (V _{BE} = -5V)	V _{CE} = 1400 V			400	μΑ
I _{EBO}	Emitter Cut-off Current (I _C = 0)	V _{EB} = 5 V			100	mA
$V_{CE(sat)^*}$	Collector-Emitter Saturation Voltage	I _C = 5 A I _B = 0.5 A			1.6	V
$V_{BE(sat)^*}$	Base-Emitter Saturation Voltage	I _C = 5 A I _B = 0.5 A			2.1	V
h _{FE} *	DC Current Gain	$I_{C} = 5 \text{ A}$ $V_{CE} = 5 \text{ V}$ $I_{C} = 5 \text{ A}$ $V_{CE} = 5 \text{ V}$ $T_{j} = 100 ^{\circ}\text{C}$	50 20			
t _s t _f	INDUCTIVE LOAD Storage Time Fall Time	$V_{CC} = 150 \text{ V}$ $I_{C} = 5 \text{ A}$ $I_{B1} = 0.5 \text{ A}$ $V_{BEoff} = -5 \text{ V}$			3 0.8	μs μs
t _s t _f	INDUCTIVE LOAD Storage Time Fall Time	$V_{CC} = 150 \text{ V}$ $I_{C} = 5 \text{ A}$ $I_{B1} = 0.5 \text{ A}$ $V_{BEoff} = -5 \text{ V}$ $T_{j} = 100 \text{ °C}$		2 0.8		μs μs
V_{F}	Diode Forward Voltage	I _F = 5 A			3	V

^{*} Pulsed: Pulse duration = 300 μs, duty cycle 1.5 %

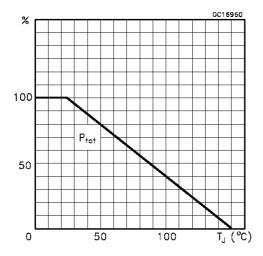
Safe Operating Area



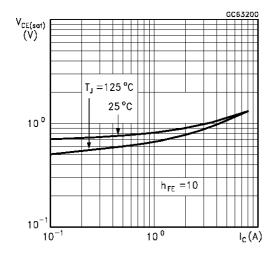
Thermal Impedance



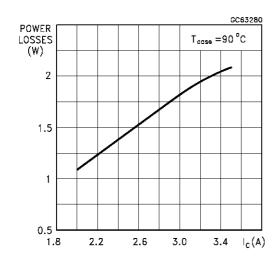
Derating Curve



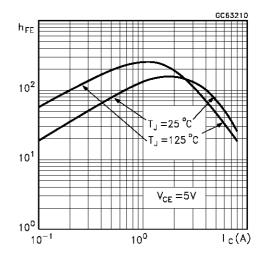
Collector Emitter Saturation Voltage



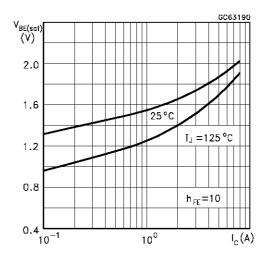
Power Losses at 16 KHz



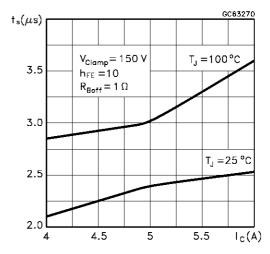
DC Current Gain



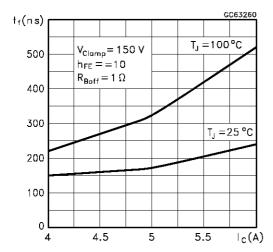
Base Emitter Saturation Voltage



Switching Time Inductive Load at 16KHz



Switching Time Inductive Load at 16KHZ

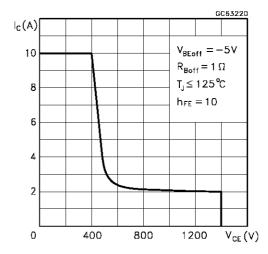


BASE DRIVE INFORMATION

In order to saturate the power switch and reduce conduction losses, adequate direct base current l_{B1} has to be provided for the lowest gain h_{FE} at 100 $^{\circ}$ C (line scan phase). On the other hand, negative base current l_{B2} must be provided to turn off the power transistor (retrace phase).

Most of the dissipation, in the deflection application, occurs at switch-off. Therefore it is essential to determine the value of I_{B2} which minimizes power losses, fall time t_f and, consequently, T_j . A new set of curves have been defined to give total power losses, t_s and t_f as a function of I_{B2} at both 16 KHz scanning frequencies for choosing the optimum negative

Reverse Biased SOA



drive. The test circuit is illustrated in figure 1.

Inductance L_1 serves to control the slope of the negative base current I_{B2} to recombine the excess carrier in the collector when base current is still present, this would avoid any tailing phenomenon in the collector current.

The values of L and C are calculated from the following equations:

$$\frac{1}{2}L(I_C)^2 = \frac{1}{2}C(V_{CEfly})^2$$
 $\omega = 2\pi f = \frac{1}{\sqrt{LC}}$

Where I_{C} = operating collector current, V_{CEfly} = flyback voltage, f= frequency of oscillation during retrace.

Figure 1: Inductive Load Switching Test Circuits.

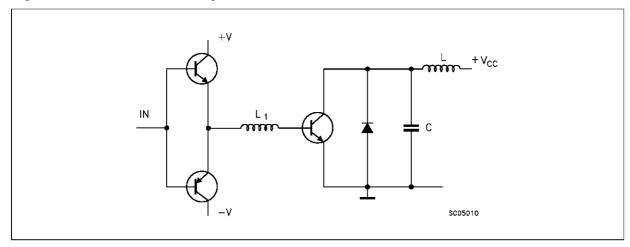
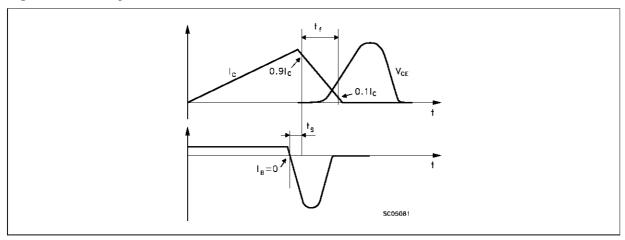
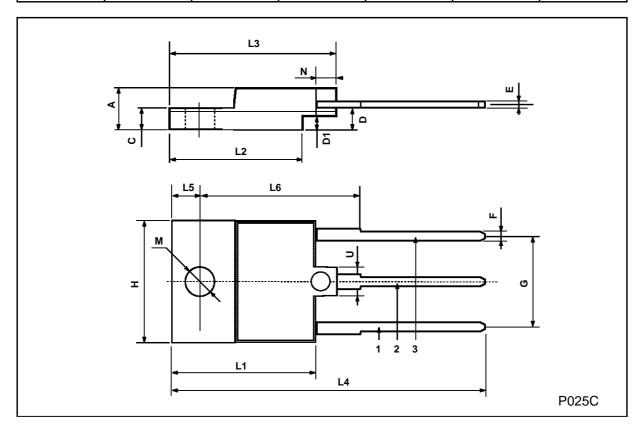


Figure 2: Switching Waveforms in a Deflection Circuit



ISOWATT218 MECHANICAL DATA

DIM.	mm		inch			
DIIVI.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
А	5.35		5.65	0.210		0.222
С	3.3		3.8	0.130		0.149
D	2.9		3.1	0.114		0.122
D1	1.88		2.08	0.074		0.081
Е	0.75		1	0.029		0.039
F	1.05		1.25	0.041		0.049
G	10.8		11.2	0.425		0.441
Н	15.8		16.2	0.622		0.637
L1	20.8		21.2	0.818		0.834
L2	19.1		19.9	0.752		0.783
L3	22.8		23.6	0.897		0.929
L4	40.5		42.5	1.594		1.673
L5	4.85		5.25	0.190		0.206
L6	20.25		20.75	0.797		0.817
М	3.5		3.7	0.137		0.145
N	2.1		2.3	0.082		0.090
U		4.6			0.181	



Information furnished is believed to be accurate and reliable. However, SGS-THOMSON Microelectronics assumes no responsability for the consequences of use of such information nor for any infringement of patents or other rights of third parties which may results from its use. No license is granted by implication or otherwise under any patent or patent rights of SGS-THOMSON Microelectronics. Specifications mentioned in this publication are subject to change without notice. This publication superseds and replaces all information previously supplied. SGS-THOMSON Microelectronics products are not authorized for use as critical components in life support devices or systems without express written approval of SGS-THOMSON Microelectonics.

 $\hbox{@ }1996\ \text{SGS-THOMSON}\ \text{Microelectronics}$ - Printed in Italy - All Rights Reserved

SGS-THOMSON Microelectronics GROUP OF COMPANIES

Australia - Brazil - Canada - China - France - Germany - Hong Kong - Italy - Japan - Korea - Malaysia - Malta - Morocco - The Netherlands - Singapore - Spain - Sweden - Switzerland - Taiwan - Thailand - United Kingdom - U.S.A

